Performance Estimation of FIR Filter using Null Convention Logic

Toushiba Khan¹, Dr. S.R.P. Sinha²

¹M.tech Scholar, Department of Electronics Engineering, Institute of Engineering and Technology
²Professor, Lucknow, Dr. APJ Abdul Kalam Technical University Lucknow India 226021

Abstract - An asynchronous delay insensitive Null Convention Logic (NCL) with dual rail signal has been used to design an improved low power high speed FIR filter and presented in this paper. NCL reduces the dynamic power consumption by reducing the switching activity. The power delay product of FIR filter using both conventional and NCL CMOS model has been estimated and compared in 90 nm technology with a frequency of 250 MHz at 1.2 V. Power reduction of 22% is achieved for the supply voltage of 2.5 V in NCL CMOS model compared to conventional CMOS model.

Key Words: Null conventional logic, asynchronous delay insensitive, Dynamic Power

1. INTRODUCTION

The development in the field of communication engineering and technology has identified the research to design low power and high speed FIR filters using the variety of concepts. The filters play a major role in the field of communication, radar, biomedical signal processing and video processing as the electronic industry is growing rapidly. Filtering is a class of signal processing, which defines the complete or partial suppression of some aspects of the signal. It removes some unwanted frequencies in order to suppress interfering signals and reduces background noise. Filters may be digital or analog filter. In digital filter, FIR filter play a vital role in the performances such as modification, reshaping and manipulation of the frequency spectrum of signal according to the desired requirements. Moreover, FIR filters are the important building blocks because of their linear phase and stability. FIR filter has a finite number of input samples which affects the generation of a given output sample and it is a time-invariant discrete linear system, frequently used in Digital Signal Processing system. It performs the frequency shaping or the linear prediction on a discrete-time input sequence \(x(0, x1, x2, \ldots)\). The output is obtained as a sum of delayed and scaled input samples. An LTI system interacts with its input signal through a process called linear convolution. The convolved sequence is given by

\[
y[n] = x[n] \ast h[n]
\]

where \(h[n]\) is the impulse response, \(x[n]\) is the input signal and \(y[n]\) is the output signal.

The FIR filter shown in Fig.1 consists of adders, multipliers and delay elements. The output can be obtained by multiplying the delayed inputs and the coefficients as given in (1).

![Figure 1 Structure of FIR filter](image)

The power dissipation of FIR filter depends up on the number of transistors in adders, multipliers and delay elements. The power dissipation is due to three sources: static power dissipation, dynamic power dissipation and short circuit power dissipation. Low power dissipation will allow the system to operate longer with the same battery. Power consumption in a logic network depends on system clock frequency \(f\), switching activity \(P\), size of transistor and their capacitance \(C_{eq}\), supply voltage \(V_{DD}\), short-circuit current \(I_{sc}\) and leakage current \(I_{le}\), as given in (3).

\[
P = \sum V_{DD} V_{sw} C_{eq} f P + V_{DD} \sum I_{sc} + V_{DD} I_{le}
\]
transistor full adder and the conventional 28-transistor CMOS adder. The disadvantage of the new adders is the threshold-voltage loss of the pass transistors. An advanced version of spurious power suppression technique (SPST) on multipliers for high-speed and low-power purposes have been proposed by Kuan-Hung Chen and Yuan-Sun Chu [2]. It leads to a 40% speed improvement. The developed SPST-equipped multiplier dissipates 0.0121 mW per MHz in 0.18μm CMOS technology and obtained 40% power reduction.

Oscal T.C. Chen et.al [3] has designed a power-efficient digital signal processor by developing a low-power adder which operates on effective dynamic data ranges. The 32-bit adder used has reduced the power dissipation in multimedia applications. Besides the ripple adder, the developed processor utilized in other adder cells, such as carry lookahead and carry-select adders to compromise complexity, speed and power consumption for application-specific integrated circuits and digital signal processors.

2- Fır filter Design Using NCL

Null Convention logic (NCL) provides an asynchronous design methodology employing dual rail signals [6], to incorporate data and control information into one mixed path. This logic is a symbolically complete logic which expresses in terms of the logic itself. NCL aims at designing VLSI devices with greater ease, with reduced power budget, lower electromagnetic interference effects and reduced noise margins. The circuits using NCL utilize dual-rail or quad-rail logic to achieve delay-insensitivity [7]–[9].

A dual-rail signal, D, consists of two wires, D0 and D1, which may assume any value from the set {DATA0, DATA1, NULL}. The DATA0 state (D0 = 1, D1 = 0) corresponds to a Boolean logic 0, the DATA1 state (D0 = 0, D1 = 1) corresponds to a Boolean logic 1, and the NULL state (D0 = 0, D1 = 0) corresponds to the empty set which means that the value of D is not yet available. The two rails are mutually exclusive, such that both rails can never be asserted simultaneously. This state is defined as an illegal state. Both dual-rail and quad-rail signals are space optimal 1-hot delay-insensitive codes, requiring two wires per bit.

NCL circuits have been constructed using threshold gates with hysteresis for its composable logic elements. One type of threshold gates is the THmn gate, where 1 ≤ m ≤ n as depicted in Fig. 2. A THmn gate corresponds to an operator with at least m signals asserted in set condition and all signals de-asserted before the output becomes asserted. Hysteresis is used to provide a means for monotonic transition and a complete transition of multi-raise signals back to NULL state before asserting the output associated with the next wave front of input data [10]. In a THmn, each of the n inputs is connected to the rounded portion of the gate. The output emanates from the pointed end of the gate. The gate’s threshold value m, is written inside of the gate. Another type of gate is weighted threshold gate denoted as THm nw1w2w3...wR in Fig. 2. Weighted threshold gates have an integer value, m ≥ wR > 1, applied to input R. Here 1 ≤ R < n; where n is the number of inputs; m is the gate’s threshold; and w1, w2, w3...wR, each > 1, are the integer weights of input1, input2, ...input R, respectively.

A- Transistor level implementation of NCL gates

NCL threshold gates are designed with hysteresis state-holding capability, such that after the output is asserted, all inputs are deasserted before the output is deasserted. NCL gates have both set and hold equations, where the set equation determines when the gate will remain asserted and the hold equation determines when the gate will remain asserted once it has been asserted. The set equation determines the gate’s functionality, where as the hold equation is the same for all the NCL gates and is simply all inputs ORed together. The general equation for an NCL gate with output Z is Z = Set + (Z - * hold), where Z- is the previous output value and Z is the new value. NCL gates can be implemented using static CMOS and semi static CMOS implementation [5]. NCL gate using CMOS technology can be implemented by complement of Z, which in general form is Z=reset + (Z - ‘* set’), where reset is complement of hold (either complement of each input, ANDed together), such that the gate is deasserted when all input are deasserted and remains deasserted while the gate’s set condition is false.

B- Basic gates using NCL

The basic gates are implemented by considering inputs “0” and “1”. For NCL implementation, boolean logic 0 is encoded as “010” and boolean logic 1 is encoded as “011”. If the input X is logic 0 then it can be encoded as X 0=1.
and \( X^1=0 \). If the input \( X \) is logic 1 then it can be encoded as \( X^0=0 \) and \( X^1=1 \).

**C. Adders using NCL**

**Half adder**: Half Adder has been designed using the EX-OR gate and AND gate which is designed using NCL gates as shown in Fig. 3.

![Fig-3 Half Adder](image)

**Full Adder**: A full adder has been designed using two half adders and an OR gate as shown in Fig. 4.

![Fig-4 Full adder](image)

**D. Multiplier using NCL**: Multiplier has been designed using the AND gates, half adders and full adders designed using NCL. D-Flip flop: D-Flip flop consists of NAND gates and inverter designed using NCL where the output is obtained during the application of high clock pulse.

**E. FIR Filter using NCL**: An 8 tap Filter has been designed using AND gates, Full adders, 16 bit Ripple carry adder, 8x8 array multiplier and D-flip flops.

**3. Results and discussion**

The 8 tap low pass FIR filter using NCL CMOS model has been designed and implemented using HSPICE and DSCH in 90nm technology.

In TH23 gate, the reset equation is \( ABC \) and the simplified set equation is \( AB+BC+AC \). Directly implementing these equations for \( Z \) and \( Z' \), after simplification, yields the static transistor-level implementation of an NCL gate as shown in Fig. 6 for TH23 gate. This requires the output to be feedback as an input to the NMOS and PMOS Logic to achieve hysteresis behavior.

**Half adder**: Half Adder has been designed as in Fig. 8 using the NCL gates and it is optimized using Threshold Combinational Reduction method. When either \( A^0 \) or \( B^0 \) is asserted \( C^0 \) will be asserted which is implemented using TH12 gate. \( C^1 \) asserts output data only when \( A^1 \) and \( B^1 \) asserts data. \( S^1 \) asserts output data when \( C^0 \) asserts data with either \( A^1 \) asserts data input value or \( B^1 \) asserts data input value but not both at same time. Hence \( S^1 \) is designed using TH33w1 gate where \( C^0 \) will be input with more weight than others \( S^0 \) asserts output data when \( C^1 \) asserts data or \( A^0 \) and \( B^0 \) asserts data input. The sum and carry is given in table 1.
Band Ci along with output from Co. $S^0 = C_0^1A_0^1 + C_0^1B_0^1 + C_1^1C_i^1 + A^1B^1C_i^1$ and $S^1 = C_0^0A_1^1 + C_0^0B_1^1 + C_0^0C_i^1 + A^1B^1C_i^1$, both are directly mapped into TH34w2 gates where Co has more weight than the other input in Table 2.

FIGURE 6: Full adder using NCL

Table 3 - NCL Full Adder

Table 3 shows the comparison of FIR filter designed using NCL logic with conventional CMOS logic. Low power and high speed has been obtained compared to conventional CMOS logic.

Table 4 Comparison of conventional CMOS with NCL logic for 8 Tap FIR filter

4- Conclusion - The 8 tap FIR Filter using conventional CMOS and NCL CMOS model have been implemented and analyzed. The FIR filter using NCL achieves high speed and low power dissipation compared with conventional CMOS model. The power delay product is better when compared with the conventional CMOS model. The power dissipation for different supply voltages and operating frequencies were estimated and found that NCL CMOS model outperform the conventional CMOS model.

REFERENCES


