Efficiency and Bandwidth Improvement Using Metamaterial of Microstrip Patch Antenna

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Abstract - This paper discusses efficiency and bandwidth improvement of microstrip patch antenna using metamaterial. A patch antenna is designed and simulated for 2.4GHz. Substrate is used for this purpose is FR4 glass epoxy with dielectric constant of 4.4, thickness of 1.6mm. The shape used for conventional patch antenna is rectangular with insert feed. For metamaterial antenna Circular Split Ring Resonators (CSRR) are used. Metamaterial CSRR is designed and simulated for 2.4GHz. After comparing both antennas bandwidth and efficiency improvement of metamaterial antenna is reported. High Frequency Structured Simulator (HFSS) EM simulator is used for this purpose.

Key Words: Microstrip Rectangular Patch antenna (MRPA), Metamaterial, Circular Split Ring Resonators (CSRR)

1. INTRODUCTION

The microstrip patch antennas are commonly used in the wireless devices. Therefore, the miniaturization of the patch antenna becomes an important issue in reducing the volume of entire communication system. The common method for reducing the microstrip patch antenna size is to utilize a high permittivity dielectric substrate [1-2]. But, these antennas are more expensive, have less radiation efficiency and narrow bandwidth. To overcome the above drawbacks, many design techniques for the patch antenna have already been proposed. These techniques are inserted slot, the meandering slots, the corrugation structure, iris structure and the shorting pin. However, all of these design strategies have limitations, such as complex structure and low performance. The design methods of the miniaturized patch antenna with metamaterial technology have been reported by some authors. Victor Georgievich Veselago was the first who theoretically presented that refractive index may also be negative. Up to this point, refractive index was traditionally regarded as having only positive values. Negative refraction can occur if both the (electric) permittivity and the magnetic permeability of a material are negative. This prediction was confirmed 33 years later when David Smith created a composite material with negative refractive index [3]. The material with negative permittivity and permeability is also known by several names such as left-handed material (LHM) and backward wave material (BWM). Although metamaterial is not present in nature, interesting properties were theoretically predicted for these substances, such as the reversal of the Snell’s Law, Doppler Effects and Cherenkov radiation and built perfect lenses [4].

By introducing the circular split ring (CSRR) structure microstrip gaps in the end behavior which is equivalent to combination of series capacitors and shunt inductors. Microstrip gap leads to negative permeability. Metamaterial antenna using six periodic CSRR gives the overall efficiency of 95.73% and return loss is -15.17 at 2.4GHz frequency [5]. In this paper, the miniaturized microstrip patch antenna with CSRRs is presented.

2. ANTENNA GEOMETRY

Figure 1 shows the geometry of the proposed conventional MSA, where a low cost glass epoxy FR4...
dielectric material with relative permittivity (εr) of 4.4 with thickness (h) of 1.6mm is chosen. The conventional MSA is designed for 2.4GHz with dimensions ‘Li’ and ‘Wi’ radiating patch, which is excited by simple 50 Ω microstrip feed having dimensions length ‘lp’ and width ‘a’ for their impedance matching

Fig -1: Conventional Patch antenna

The width of the patch antenna is calculated by formula

\[ W = \frac{c}{2f \sqrt{(\varepsilon_r+5)\varepsilon_r}} \]  

(1)

Where

C - Velocity of light in air = 3 \times 10^8 m/s.
F - Frequency of antenna = 1.4 \times 10^9 Hz.
εr - Dielectric constant of substrate = 4.4

\[ W = 39.03 \text{ mm.} \]

**Effective dielectric constant** as given by Hammersted and Jensen [6],

\[ \varepsilon_{\text{eff}} = \frac{\varepsilon_r + 1}{2} + \frac{\varepsilon_r - 1}{2} \left[ 1 + 12 \frac{h}{W} \right]^{-1/2} \]  

(2)

If the height of the substrate = 1.6 mm, then from equation (2)

\[ \varepsilon_{\text{eff}} = 4.2. \]

Due to effect of fringing fields the length (L) increases by ΔL from physical length, which is given by length extension (ΔL)

\[ \Delta L = 0.412 h \frac{(\varepsilon_{\text{eff}}+5.2)(\varepsilon_{\text{eff}}+0.264)}{(\varepsilon_{\text{eff}}-0.3)\varepsilon_{\text{eff}}+0.3} \]  

(3)

As patch width is 39.03mm and ε_{\text{eff}} = 4.2 then from eq. (3)

\[ \Delta L = 0.0024 \]

**Effective length of the patch**

\[ L_{\text{eff}} = \frac{c}{2f \sqrt{(\varepsilon_{\text{eff}}+5)\varepsilon_{\text{eff}}}} \]  

(4)

\[ L_{\text{eff}} = 30.49 \text{ mm.} \]

**Actual length of the patch:**

\[ L = L_{\text{eff}} - 2\Delta L \]  

(5)

\[ L = 30 \text{ mm.} \]

**Ground plane calculations:**

\[ 6h+W = 49.83 = 50\text{mm} \]  

(6)

\[ 6h+L = 39.04 = 40\text{mm} \]  

(7)

The dimensions of conventional patch antenna obtained using the above equations are listed in Table 1.

**Table 1:** Parameters of Conventional Patch Antenna

<table>
<thead>
<tr>
<th>Input parameters</th>
<th>Values</th>
</tr>
</thead>
<tbody>
<tr>
<td>Width of patch</td>
<td>40 mm</td>
</tr>
<tr>
<td>Effective dielectric constant</td>
<td>4.2</td>
</tr>
<tr>
<td>Length of patch</td>
<td>30 mm</td>
</tr>
<tr>
<td>Conductance of the patch (G1)</td>
<td>1.0816 \times 10^{-2}</td>
</tr>
<tr>
<td>Insert feed point distance (y_0)</td>
<td>13 mm</td>
</tr>
<tr>
<td>Width of the strip line (W)</td>
<td>4.3 mm</td>
</tr>
<tr>
<td>Substrate length and width (WxL)</td>
<td>50mm x 40mm</td>
</tr>
</tbody>
</table>
3. METAMATERIAL GEOMETRY

After obtaining the results for conventional patch, a metamaterial structure is incorporated into the antenna design on another layer of FR4 substrate of thickness 1.6mm. A small circular split ring resonator (SRR) loaded microstrip patch antenna with a CPW fed is designed. The circular split ring resonator is designed and tested in a waveguide.

Figure 3 shows the configuration of the proposed antenna.

For the observation of the proposed metamaterial structure of figure 2, Dominant mode waveguide having lowest cut-off frequency in TE10 mode have been designed. The cutoff frequency is the frequency below which the waveguide will not operate. Cut-off frequency for waveguide 1.8 GHz which calculated from the following formula. Proposed waveguide structure shown in figure 3(b).

The cutoff frequency is given by,

\[ f_{\text{cut}} = \frac{c}{2L} \sqrt{\left(\frac{\pi}{a}\right)^2 + \left(\frac{\pi}{b}\right)^2} \]

(a)

\[ a = 83.33 \times 10^{-3} = 83.33 \text{ mm} \]

(b)

\[ b = \frac{a}{2} = 41.66 \times 10^{-3} = 41.66 \text{ mm} \]

As per calculations we got cut-off frequency for waveguide to 1.8276 GHz which shown in figure 4. In order to know for which frequency the metamaterial structure resonates. The proposed metamaterial structure is placed between two waveguide ports as shown in figure 5. Left and Right hand side of the y-axis and z-axis are defined as Perfect Electric (PEC). Two x-planes given as source port.

![Fig -2: Proposed CSRR Structure](image)

![Fig -3: (a) Geometry for rectangular waveguide, (b) waveguide structure for cut-off Frequency of 1.8GHz](image)

![Fig -4: Cut-off frequency of the waveguide to 1.8 GHz](image)
The S-parameters for the proposed metamaterial structures are observed. $S_{-11}$ is $-26$ dB at 2.4GHz is reported as in figure 6. The designed metamaterial loaded patch antenna with its E-Field is reported in figure 7.

Proposed patch antenna has been designed by loading with metamaterial on U-shape. For providing metamaterial property (interdigital capacitance and inductance) in the proposed antenna CSRR's are introduced to both arms of U-shape patch antenna. Split ring resonators don’t have any physical connection. So CSRR’s are excited by Electromagnetic coupling between ring and source patch. Figure 7 shows the current distribution on the patch near the resonance at 2.4 GHz. At the upper edges of U-shape maximum E-Field is observed. The charges at upper edges will jump to CSRR and these rings will get excited. The current is distributed mainly along edges and the feed point.

Fig -5: Proposed CSRR structure is placed between two waveguide ports.

Fig -6: $S_{11}$ characteristics of metamaterial structure

4. Results and Analysis
1. $S_{11}$-characteristics:

Fig -7: Metamaterial Loaded Patch Antenna and its E-Field

Fig -8: (a) Simulation of Return Loss $S_{11}$ of conventional patch antenna. (b) $S_{11}$ Simulation of Return Loss $S_{11}$ of metamaterial loaded patch antenna.
In conventional patch RL = -19 dB at resonating frequency of 2.4 GHz is reported as in figure 8(a) and the bandwidth achieved is only 16 MHz. In metamaterial loaded patch antenna RL = -15.17 dB at resonating frequency of 2.4 GHz is reported. The return loss of -10 dB is observed over the range of 1.6 GHz-2.7 GHz i.e. in the proposed metamaterial structure achieved bandwidth is up to 1GHz seen from figure 8(b).

2. Radiation Pattern:

Figure 9, is the simulated radiation pattern at frequency for 2.4 GHz. From figure 9(b), it is clear that the radiation pattern displays a directional behavior, with its main lobe direction at 0˚ and 180˚.

![Radiation Pattern 1](image1)

![Radiation Pattern 2](image2)

Fig -9: (a) Radiation Pattern Conventional Patch Antenna. (b) Radiation pattern of metamaterial loaded antenna.

Figure 9(b), this indicates that the concentration of the field focuses on the sides of the patch. Meanwhile, the lobes are suppressed at 90˚ and 270˚, which originate from the front part and back part of the patch antenna.

Radiation efficiency is also measured, for conventional patch we got efficiency of 56% only. While for metamaterial loaded patch efficiency is approximately 96%, around 40% efficiency increment is achieved.

3. VSWR Characteristics:

![VSWR Characteristic](image3)

Fig -10: (a) VSWR characteristic for conventional patch, (b) VSWR characteristic for metamaterial loaded patch.

![Far field gain](image4)

Fig -11: Far field gain of the conventional patch antenna and Metamaterial loaded patch antenna.
Table -2: Comparison Table between Conventional Patch antenna and Metamaterial Loaded Patch antenna

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Simple Patch</th>
<th>Metamaterial patch</th>
</tr>
</thead>
<tbody>
<tr>
<td>Size (in mm)</td>
<td>50x40x1.6</td>
<td>50x40x1.6</td>
</tr>
<tr>
<td>Frequency</td>
<td>2.4 GHz</td>
<td>2.4 GHz</td>
</tr>
<tr>
<td>Frequency range</td>
<td>2.32 GHz - 2.45 GHz</td>
<td>1.62 GHz - 2.7 GHz</td>
</tr>
<tr>
<td>Bandwidth achieved</td>
<td>11 MHz</td>
<td>1 GHz</td>
</tr>
<tr>
<td>Return Loss</td>
<td>-19 dB</td>
<td>-15.17 dB</td>
</tr>
<tr>
<td>Radiation Pattern</td>
<td>One-directional</td>
<td>Bi-directional</td>
</tr>
<tr>
<td>Efficiency</td>
<td>56.96%</td>
<td>95.73%</td>
</tr>
</tbody>
</table>

5. CONCLUSION

The effect of periodic structure of SRRs on a microstrip patch has been studied. The bandwidth and efficiency of the antenna showed an increase as compared to normal patch. Bandwidth achievement is 1GHz as compared to 16MHz for the conventional patch antenna. The efficiency of metamaterial CSRR is 95.73% as compared to 56.96% of conventional patch. This is due to introduction of interdigital capacitance and inductance.

REFERENCES


BIOGRAPHIES

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