GPS MICRO STRIP RECTANGULAR PATCH ANTENNA AT 2.45 GHz MOUNTED OVER ANGULAR TOWERS

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Abstract—Antennas are mounted over Telecommunication Towers where Signaling is done at the cell site. Thus designing of such an antenna used in telecom unit is of great interest. A GPS Receiver Micro-Strip Patch antenna over CST Microwave Studio is designed. All GPS receivers use an antenna to receive these signals. The signals from the GPS satellites also operate in the C band the antenna must have a clear view of the sky and thus be mounted on a roof, or in some cases in a window. The antennas are relatively small, coffee cup size or smaller, and are connected to the GPS receiver typically via coaxial cable. Since the GPS signal is very weak the antenna usually amplifies the signal to drive it through the cable to the receiver. Thus for designing purpose the patch antenna is used which in demand these days because of its small size and low cost.

I. INTRODUCTION

Antenna is one of the main components for GPS receiver. A good antenna design can reduce system requirements and improve overall system performance. The antennas are classified on the basis of different factors. The most convenient and important factor for classification and designing of antenna is the frequency. Since the antenna designed is to be used in GPS system, it was designed for 2.45MHz.

A Micro strip Antenna is also termed as patch antenna. Patch antenna are becoming increasingly useful because they can be printed directly onto a circuit board. Micro strip antennas are becoming very widespread within the mobile phone market, as these are low cost, have a low profile and are easily fabricated. The basic micro strip patch antenna has a very thin conductor on one side of the substrate and the other side has a conductive ground plane. In order to simplify analysis and performance prediction, the patch is taken to be rectangular.

Micro strip patch antennas can be fed by a variety of methods. These methods can be classified into two categories- contacting and non-contacting. In the contacting method, the RF power is fed directly to the radiating patch using a connecting element such as a micro strip line. In the non-contacting scheme, electromagnetic field coupling is done to transfer power between the micro strip line and the radiating patch. The four most popular feed techniques used are the micro strip line, coaxial probe (both contacting schemes), aperture coupling and proximity coupling (both non-contacting schemes). For more spurious feed radiation, better reliability, easy fabrication and easy impedance matching micro strip line feed method is opted in this design.

II. ANTENNA DESIGN

A. Design of Rectangular patch

For a rectangular patch, the width W and the length L are:

$$W = \frac{C}{2f_0} \left( \frac{\varepsilon_r + 1}{2} \right)^{0.5}$$

(3.1)

$$L = \frac{c}{2f_0\varepsilon_{eff}} = 2\Delta l$$

(3.2)

$$\varepsilon_{eff} = \frac{\varepsilon_r + 1}{2} + \frac{\varepsilon_r - 1}{2} \left( \frac{1}{1 + \frac{12\Delta h}{W}} \right)^{0.5}$$

(3.3)

$$\Delta l = 0.412 \left[ \frac{0.262 + (W/h)}{0.813 + (W/h)} \right] \left( \frac{\varepsilon_{eff} + 0.3}{\varepsilon_{eff} - 0.258} \right)$$

(3.4)

Where,

c = Velocity of light = 3 X 10\(^{11}\) mm/second.
fr = The operating frequency in GHz
\varepsilon_r = Dielectric constant of the substrate
\varepsilon_{eff} = Effective dielectric constant of the substrate
\Delta l = the line extension in mm
h = Thickness of the substrate in mm
B. Design of 50 ohm feed line

Different values of W/d ratio are:

$$W = \left[ \frac{8e^{\varepsilon_r}}{\varepsilon^{2\Delta A}} - 2 \right]$$

For W/d < 2

$$W = \left[ \frac{1}{2\sqrt{B - 1 - \ln\left(2B - 1\right) + \left(\frac{\varepsilon_r - 1}{2\varepsilon_r}\right) \ln\left(2B - 1\right) + 0.39 \left(\frac{0.61}{\varepsilon_r}\right)} \right]$$

W/d > 2

(3.5)

Where

$$A = \left( \frac{Z_0}{60} \right) \left( \frac{\sqrt{\varepsilon_r + 1}}{2} + \frac{\varepsilon_r - 1}{\varepsilon_r + 1} \right) \left( 0.23 + \frac{0.11}{\varepsilon_r} \right)$$

(3.6)
B = \left[ \frac{377 \Pi}{2Z_0 \sqrt{\epsilon_r}} \right] \quad (3.7)
\epsilon_{re} = \frac{\epsilon_r + 1}{2} + \left( \frac{\epsilon_r - 1}{2} \right) \left( \frac{1}{\sqrt{1 + \frac{12H}{w}}} \right) \quad (3.8)
\lambda_{g} = \frac{\lambda_0}{\sqrt{\epsilon_{re}}} \quad (3.9)
\lambda_{eff} = \frac{\lambda_{g}}{4}

C. Designing of λ/4 Transformer Line

\lambda/4 transformer line is designed to precisely match the impedance between input feed line and the patch. The input resistance to patch is given by \( R_{in} \):

\[ R_{in} = \frac{120\lambda_{eff}}{2W_{patch}} \quad (3.11) \]

\[ Z = \sqrt{50R_{in}} \quad (3.12) \]

The same formulas which are used for calculating length and width of 50 ohm line have been used for calculating length and width of transformer line.

II. DESIGN PARAMETERS

The rectangular patch is realized with a substrate material of GML 1000 and permittivity of 3.2.

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Designed values</th>
<th>Optimized values</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operating frequency</td>
<td>2.45</td>
<td>2.45</td>
</tr>
<tr>
<td>Patch length (AL)</td>
<td>34.046</td>
<td>32.414</td>
</tr>
<tr>
<td>Patch width (AW)</td>
<td>42.248</td>
<td>39.2</td>
</tr>
<tr>
<td>Quarter line length (QL)</td>
<td>20.634</td>
<td>19.2</td>
</tr>
<tr>
<td>Quarter line width (QW)</td>
<td>0.9255</td>
<td>0.5478</td>
</tr>
<tr>
<td>Substrate Length (SL)</td>
<td>70</td>
<td>70</td>
</tr>
<tr>
<td>Substrate width (SW)</td>
<td>70</td>
<td>70</td>
</tr>
<tr>
<td>Input Line Length</td>
<td>19.173</td>
<td>5</td>
</tr>
<tr>
<td>Input Line Width</td>
<td>1.833</td>
<td>1.833</td>
</tr>
</tbody>
</table>

Table I. Design Parameters

III. SIMULATION RESULTS AND DISCUSSIONS

A. Return loss:

As expected, Operating frequency of 2.45GHz is obtained as shown in the figure 2 and 3. Referring to figure 3 the red curve shows the return loss of initial structure designed and green curve shows the return loss of optimized structure. Both have a peak at 2.45 GHz but the difference lies in the accuracy.

The value of S11 for initial structure was -14.678371 dB which was improved to the best value of -29.822192 dB after optimization of the structure.

Fig 1: Structure of Rectangular Patch Antenna

Fig 2 Return Loss
B. Result of variation of antenna patch length $AL$:

A parametric study was done wherein, antenna patch length ($AL$) was varied and the value of return loss was plotted in the graph as shown in figure 4 which depicts that the Variation in value of $AL$ is also reflected in the value of operating frequency. Referring to figure 4 it is observed that the antenna patch length ($AL$) is inversely proportional to the operating frequency. As the value of $AL$ increases from 32.414mm to 35.217mm the operating frequency decreases from 2.45GHz. Hence, it is observed that the red curve has a precise value of 2.45 GHz so the optimized value with $AL$ of 32.414mm as shown in table 1.

C. Result of variation of antenna patch width $AW$:

A parametric study was done wherein, antenna patch width ($AW$) was varied and the value of return loss was plotted in the graph as shown in figure 5 which depicts that the variation in value of $AW$ is also reflected in the value of return loss which can be clearly seen in figure 6.

D. Result of variation of quarter line width $QW$:

A similar study was done wherein, quarter line width ($QW$) was varied and the value of return loss was plotted in the graph as shown in figure 7 which depicts that the variation in value of $QW$ is also reflected in the value of return loss which can be clearly seen in figure 8 and thus precisely matches the impedance between patch and input line.
E. Final return loss after all optimizations:

The graph after optimization of the structure is shown in figure 9 that concludes the value of S11 Parameter to be -29.822192dB at the frequency of 2.45 GHz which is the required operating frequency.

F. Efficiency curve:

The above figure 10 depicts the radiation efficiency of 66.9% and total efficiency of 66.83% at the operating frequency of 2.45 GHz.

G. Energy Balance curve:

From figure 11 it is depicted that as soon as the simulator achieved the accuracy of -30dB the simulation stopped and it took approximately 7ns.

H. Magnetic Field:

Figure 12 shows the magnetic field variation in the Patch at frequency of 2.45GHz.

I. Electric Field:

Similarly figure 13 sows the electric field variation at operating frequency. It would be interesting to note that cutting a slot at a point where the magnitude of E field is high, disturbs the pattern of Electric field the most.

J. Surface current:

Figure 14 shows the pattern of surface current over the patch.
K. 3D Far Field:

Figure 15 shows the Radiation Pattern of the patch antenna designed where red portion represents the maximum value of gain or directivity. So the line of maximum directivity is along z axis according to the patch designed in x-y plane.

Figure 16 depicts the major lobe of patch antenna along z axis and minor lobe is along -z axis where gain or directivity will be minimum and is represented by blue color. Thus similar patterns are expected from GPS antennas.

L. E and H Plane Polar Plot:

Theta is the angle between the x axis and the line from the origin to the point of measurement. Phi is the angle between the z axis and the line joining the origin to the point of measurement. Varying phi while keeping theta constant gives H-Plane polar plot whereas varying theta while keeping phi constant gives the E-Plane Polar Plot.

Figure 17 represents the H Plane polar plot where theta is fixed at 0 degree and phi is varied from 0 to 360 degrees giving a constant gain of 7.1dB in all directions.

Figure 18 represents the H Plane polar plot where theta is fixed at 90 deg. and phi is varied from 0 to 360 deg. giving a major lobe at 307 deg. and at 233 deg. having the magnitude of -2.9dB.

Figure 19 represents the E Plane polar plot where phi is fixed at 0 deg. and theta is varied from 0 to 360 deg. giving a major lobe at 0 deg. having the gain of 7.1dB.
Figure 20 represents the E Plane polar plot where phi is fixed at 90 deg, and theta is varied from 0 to 360 deg, giving a major lobe at 0 deg, having the gain of 7.1dB.

IV. FABRICATION RESULTS AND DISCUSSIONS

Figure 21 represents the return loss of the fabricated antenna with the help of Vector Network Analyzer. These results show that the operating frequency of fabricated antenna is 2.4618GHz which is almost near to the value of frequency the antenna was designed for.

The value of Return Loss (S11) in the graph is -11.87dB which is more than 99.9% of accuracy, hence the fabricated antenna is reliable for real time applications.

Figure 22 is the polar plot of radiation pattern of the fabricated antenna as obtained by the measurement done in Anechoic Chamber whose shape and direction of radiation are as expected.

V. CONCLUSIONS

For achieving the design of the GPS antenna, software used is CST Microwave Studio where the specification of operating frequency and substrate material used for microstrip patch antenna was initially selected to be 2.45GHz and GML1000 respectively.

As expected, Operating frequency of 2.45GHz is obtained where the return loss of structure designed and the return loss of optimized structure both have a peak at 2.45 GHz but the difference lies in their magnitude. A parametric study was done where, antenna patch length (AL), antenna patch width (AW) and Quarter wave line width (QW) of the designed antenna is varied. This variation tells that the operating frequency is inversely proportional to the AL. Also the results show that the increase in the value of AW improves the value of return loss and thus optimization is achieved. Variation in QW helps in precisely matching the impedance between patch and input line. Next energy balance curve of the antenna was observed where it was plotted that as soon as radiated power was matched with the energy of input feed line to an accuracy of -30dB, the simulator stopped. Radiation Pattern of the patch.
The antenna designed shows that the maximum value of gain or directivity is along z axis and minimum is along –z axis.

Vector network analyzer was used to check the parameters of the fabricated antenna and it was observed that the operating frequency of 2.45GHz was achieved with the value of accuracy of -11dB which is suitable for real time applications. Further the antenna was tested in anechoic chamber and the shape and direction of radiation pattern plotted was as expected.

VI. FUTURE PROSPECTS

Different types of substrate material other than GML1000 and background material other than Vacuum can be chosen and the best combination of two can be found out for best results. The accuracy can be further improved for the designed structure. In the work done the accuracy maintained was -30dB which can be reduced to improve the performance. Similarly structure can be further optimized for precise results. The optimized accuracy was -30 dB but the fabricated antenna accuracy came to be -11 dB so the fabrication process must be improved to decrease the deterioration in the results. For simulation of antenna designed the transient solver was used. Outputs of frequency solver can also be simulated and compared for further analysis. The similar studies can be done for antennas working in frequency bands other than C band (2.45GHz).

REFERENCES


[3] Norfishah Ab Wahab, Zulkifli Bin Maslan, Wan Norasyafizan W. Muhamad, Norhayati Hamzah "Microstrip Rectangular 4x1 Patch Array Antenna at 2.5GHz for WiMax Application" 2010 Second International Conference on Computational Intelligence, Communication Systems and Networks


